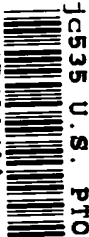


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PATENT APPLICATION
TRANSMITTAL**

(Only for new nonprovisional applications under 37 C.F.R. § 1.53(b))

Attorney Docket No. 203-2391 CON (2060 CON)

First Inventor or Application Identifier Daniel J. Becker

Title ELECTROSURGICAL GENERATOR POWER CONTROL..

Express Mail Label No. EM401294151US

APPLICATION ELEMENTS

See MPEP chapter 600 concerning utility patent application contents.

ADDRESS TO: Assistant Commissioner for Patents
Box Patent Application
Washington, DC 20231

1. ☒ Fee Transmittal Form (e.g., PTO/SB/17)
(Submit an original and a duplicate for fee processing)
2. ☒ Specification [Total Pages 23]
(preferred arrangement set forth below)
- Descriptive title of the invention
 - Cross References to Related Applications
 - Statement Regarding Fed sponsored R & D
 - Reference to Microfiche Appendix
 - Background of the invention
 - Brief Summary of the invention
 - Brief Description of the Drawings (if filed)
 - Detailed Description
 - Claim(s)
 - Abstract of the Disclosure
3. ☒ Drawing(s) (35 U.S.C. 113) [Total Sheets 3]
4. Oath or Declaration [Total Pages 4]
- a. ☐ Newly executed (original or copy)
- b. ☒ Copy from a prior application (37 C.F.R. § 1.63(d))
(for continuation/divisional with Box 17 completed)
[Note Box 5 below]
- i. ☐ **DELETION OF INVENTOR(S)**
Signed statement attached deleting inventor(s) named in the prior application, see 37 C.F.R. §§ 1.63(d)(2) and 1.33(b).
5. ☐ Incorporation By Reference (useable if Box 4b is checked)
The entire disclosure of the prior application, from which a copy of the oath or declaration is supplied under Box 4b, is considered to be part of the disclosure of the accompanying application and is hereby incorporated by reference therein.

6. ☐ Microfiche Computer Program (Appendix)
7. Nucleotide and/or Amino Acid Sequence Submission (if applicable, all necessary)
- a. ☐ Computer Readable Copy
- b. ☐ Paper Copy (identical to computer copy)
- c. ☐ Statement verifying identity of above copies

ACCOMPANYING APPLICATION PARTS

8. ☐ Assignment Papers (cover sheet & document(s))
9. ☐ 37 C.F.R. § 3.73(b) Statement (when there is an assignee) ☐ Power of Attorney
10. ☐ English Translation Document (if applicable)
11. ☐ Information Disclosure Statement (IDS)/PTO-1449 ☐ Copies of IDS Citations
12. ☐ Preliminary Amendment
13. ☒ Return Receipt Postcard (MPEP 503)
(Should be specifically itemized)
14. ☐ Statement(s) ☐ Statement filed in prior application, Status still proper and desired (PTO/SB/09-12)
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17. If a CONTINUING APPLICATION, check appropriate box, and supply the requisite information below and in a preliminary amendment:

☒ Continuation ☐ Divisional ☐ Continuation-in-part (CIP) of prior application No: 08, 533,891

Prior application information: Examiner Ryan Carter Group / Art Unit: 3311

18. CORRESPONDENCE ADDRESS

☐ Customer Number or Bar Code Label or ☒ Correspondence address below
(Insert Customer No. or Attach bar code label here)

Name	John C. Andres, Esq.				
	United States Surgical Corporation				
Address	150 Glover Avenue				
City	Norwalk	State	Connecticut	Zip Code	06856
Country	U.S.A.	Telephone	203-845- 4018	Fax	203-846-5988

Name (Print/Type)	John G. Tutunjian	Registration No. (Attorney/Agent)	39,405
Signature		Date	May 21, 1998

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These are the fees effective October 1, 1997.
Small Entity payments must be supported by a small entity statement,
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See 37 C.F.R. §§ 1.27 and 1.28.

TOTAL AMOUNT OF PAYMENT (\$ 790.00

Complete if Known

Application Number	
Filing Date	
First Named Inventor	Daniel J. Becker
Examiner Name	Ryan Carter
Group / Art Unit	3311
Attorney Docket No.	203-2091 CON (2060 CON)

METHOD OF PAYMENT (check one)

- 1.
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- The Commissioner is hereby authorized to charge indicated fees and credit any over payments to:

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Deposit Account Name Dilworth & Barrese☒ Charge Any Additional Fee Required Under 37 C.F.R. §§ 1.16 and 1.17 ☐ Charge the Issue Fee Set in 37 C.F.R. § 1.18 at the Mailing of the Notice of Allowance

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FEE CALCULATION**1. BASIC FILING FEE**

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description	Fee Paid
101 790	201 395	Utility filing fee	790.00
106 330	206 165	Design filing fee	
107 540	207 270	Plant filing fee	
108 790	208 395	Reissue filing fee	
114 150	214 75	Provisional filing fee	
SUBTOTAL (1)			(\$ 790.00)

2. EXTRA CLAIM FEES

Total Claims	Extra Claims	Fee from below	Fee Paid
9	-20** = 0	X	0
3	-3** = 0	X	0
Multiple Dependent			0

**or number previously paid, if greater; For Reissues, see below

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description
103 22	203 11	Claims in excess of 20
102 82	202 41	Independent claims in excess of 3
104 270	204 135	Multiple dependent claim, if not paid
109 82	209 41	** Reissue independent claims over original patent
110 22	210 11	** Reissue claims in excess of 20 and over original patent

SUBTOTAL (2) (\$ 0

FEE CALCULATION (continued)**3. ADDITIONAL FEES**

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description	Fee Paid
105 130	205 65	Surcharge - late filing fee or oath	
127 50	227 25	Surcharge - late provisional filing fee or cover sheet	
139 130	139 130	Non-English specification	
147 2,520	147 2,520	For filing a request for reexamination	
112 920*	112 920*	Requesting publication of SIR prior to Examiner action	
113 1,840*	113 1,840*	Requesting publication of SIR after Examiner action	
115 110	215 55	Extension for reply within first month	
116 400	216 200	Extension for reply within second month	
117 950	217 475	Extension for reply within third month	
118 1,510	218 755	Extension for reply within fourth month	
128 2,050	228 1,030	Extension for reply within fifth month	
119 310	219 155	Notice of Appeal	
120 310	220 155	Filing a brief in support of an appeal	
121 270	221 135	Request for oral hearing	
138 1,510	138 1,510	Petition to institute a public use proceeding	
140 110	240 55	Petition to revive - unavoidable	
141 1,320	241 660	Petition to revive - unintentional	
142 1,320	242 660	Utility issue fee (or reissue)	
143 450	243 225	Design issue fee	
144 670	244 335	Plant issue fee	
122 130	122 130	Petitions to the Commissioner	
123 50	123 50	Petitions related to provisional applications	
126 240	126 240	Submission of Information Disclosure Stmt	
581 40	581 40	Recording each patent assignment per property (times number of properties)	
146 790	246 395	Filing a submission after final rejection (37 CFR 1.129(a))	
149 790	249 395	For each additional invention to be examined (37 CFR 1.129(b))	
Other fee (specify) _____			
Other fee (specify) _____			
* Reduced by Basic Filing Fee Paid			
SUBTOTAL (3)			(\$ 0

SUBMITTED BY

Typed or Printed Name John G. Tutunjian

Signature

Date 5/21/98

Complete (if applicable)

Reg. Number 39,405

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ELECTROSURGICAL GENERATOR POWER CONTROL CIRCUIT AND METHOD

Related Application Information

This application is a Continuation application of U.S. Patent Application Serial No. 08/533,891 filed on September 26, 1995, pending.

5 1. **Field of the Invention** A constant power control circuit for an electrosurgical generator and a method for maintaining the electrical power output of an electrosurgical generator at a generally constant level throughout a given tissue impedance range.

10 2. **Background of the Disclosure** An electrosurgical generator is used in surgical procedures to deliver electrical energy to the tissue of a patient. An electrosurgical generator often includes a radio frequency generator and its controls. When an electrode is connected to the generator, the electrode can be used for cutting or coagulating the tissue of a patient with high frequency electrical energy. During normal operation, alternating electrical current from the generator flows between an active electrode and a return electrode by passing through the tissue and bodily fluids of a patient.

15 The electrical energy usually has its waveform shaped to enhance its ability to cut or coagulate tissue. Different waveforms correspond to different modes of operation of the generator, and each mode gives the surgeon various operating advantage. Modes may include cut, coagulate, a blend thereof, desiccate, or spray. A surgeon can easily select and change the different modes of operation as the surgical procedure progresses.

20 In each mode of operation, it is important to regulate the electrosurgical power delivered to the patient to achieve the desired surgical effect. Applying more electrosurgical power than necessary results in tissue destruction and prolongs healing. Applying less than the desired amount of electrosurgical power inhibits the surgical procedure. Thus, it is desirable to control the output energy from the electrosurgical generator for the type of tissue being treated.

25 Different types of tissues will be encountered as the surgical procedure progresses and each unique tissue requires more or less power as a function of frequently changing tissue impedance. Even the same tissue will present a different load impedance as the tissue is desiccated.

Two conventional types of power regulation are used in commercial electrosurgical generators. The most common type controls the DC power supply of the generator by limiting the amount of power provided from the AC mains to which the generator is connected. A feedback control loop regulates output voltage by comparing a desired voltage with the output voltage supplied by the power supply. Another type of power regulation in commercial electrosurgical generators controls the gain of the high-frequency or radio frequency amplifier. A feedback control loop compares the output power supplied from the RF amplifier for adjustment to a desired power level. Generators that have feedback control are typically designed to hold a constant output voltage, and not to hold a constant output power.

U.S. Patents 3,964,487; 3,980,085; 4,188,927 and 4,092,986 have circuitry to reduce the output current in accordance with increasing load impedance. In those patents, constant voltage output is maintained and the current is decreased with increasing load impedance.

U.S. Patent 4,126,137 controls the power amplifier of the electrosurgical unit in accord with a non linear compensation circuit applied to a feedback signal derived from a comparison of the power level reference signal and the mathematical product of two signals including sensed current and voltage in the unit.

U.S. Patent 4,658,819 has an electrosurgical generator which has a microprocessor controller based means for decreasing the output power as a function of changes in tissue impedance.

U.S. Patent 4,727,874 includes an electrosurgical generator with a high frequency pulse width modulated feedback power control wherein each cycle of the generator is regulated in power content by modulating the width of the driving energy pulses.

U.S. Patent 3,601,126 has an electrosurgical generator having a feedback circuit that attempts to maintain the output current at a constant amplitude over a wide range of tissue impedances.

None of the aforementioned U.S. Patents include a constant power control circuit that provides for a generally constant output power while also providing a linear adjustment to account for the unique waveform crest factors associated with different operational modes.

The preferred constant power control circuit and method provided herein allows for output power control by way of a unique and simple linear conversion circuit coupled with protection circuitry that prevents the electrosurgical generator from being over-driven during high and/or low impedance loading. The preferred
5 constant power control circuit also reduces the severity of exit sparking by responding quickly to high impedance indications while nonetheless maintaining substantially increased power levels throughout a predetermined patient tissue impedance range.

10

SUMMARY OF THE INVENTION

A constant power control circuit for use with an electrosurgical generator. The constant power control circuit and method may be included as an integral part of the overall electrosurgical generator's circuitry, or may be designed as a separate
15 unit that connects to, and controls, an electrosurgical generator. The constant power control circuit and method may be embodied through a variety of analog and/or digital circuit components or arrangements, including software running on computational and memory circuitry.

The constant power control circuit and method maintain the output power of
20 the electrosurgical current at a generally constant level over a finite patient tissue impedance range. The preferred patient tissue impedance range is about 300 to 2500 ohms.

The constant power control circuit and method provide the capability to control the output power of the electrosurgical generator without having to actually
25 monitor the amplitude of both the output current and output voltage. This allows for a simple constant power control circuit and method which operate to control the power output without having to calculate the actual power output of the electrosurgical generator.

While the constant power control circuit may be used to control
30 electrosurgical generators of varying designs, it is preferred that the electrosurgical generator includes a power selection system wherein the user may initialize, set, monitor, and/or control the operation of the electrosurgical generator. It is also preferred that the power selection system produces a control voltage signal that acts

to control a high voltage direct current supply which in turn acts to supply a high voltage signal to an output switching radio frequency stage. Then output switching radio frequency stage creates an electrosurgical energy between two output electrodes. The preferred electrosurgical generator need not be limited to these three
 5 functional elements, for example the electrosurgical generator could also include additional safety, monitoring, signal modification/conditioning, and/or feedback circuitry or functional elements/processes. The actual electrosurgical generator's design may include the use of digital components and signaling and/or analogue components and signaling, or may be embodied, completely or partially within a
 10 software process running on hardware components.

The constant power control circuit includes a current sampling circuit, a linear conversion circuit, and a feedback correction circuit. The current sampling circuit is coupled to one of the output electrodes, and functions so as to produce a sampled current signal that is proportional to the average current flowing through the output
 15 electrode.

The linear conversion circuit which is connected to the current sampling circuit internally generates one or more multiplier reference signals and one or more offset reference signals, each of which is used to modify the sampled current signal in accord with the crest factor associated with the electrosurgical energy output by
 20 the electrosurgical generator; the modified signal being a linear converted signal.

The feedback correction circuit which is electrically connected to receive the linear converted signal from the linear conversion circuit and the control voltage signal from the power selection system functions to produce a feedback control signal which it then supplies to the power selection system, within the
 25 electrosurgical generator, so as to cause the power selection system to control the amount of electrosurgical energy created. The feedback correction circuit functions so as to determine the difference in amplitude between the control voltage signal and the linear converted signal and to then add this difference to the control voltage signal to produce a feedback control signal. The feedback correction circuit may also
 30 be connected to the primary transformer winding within the output switching radio frequency stage, or its equivalent, thereby allowing the feedback correction circuit to detect high impedance loading between the output electrodes and to reduce the amplitude of the feedback control signal to protect the circuitry and/or the patient

from excessive current and/or voltage levels. A high impedance load is generally considered to be above 2500 ohms. The feedback correction circuit may also include circuitry or processes that substitute another signal for the feedback control signal when the impedance loading between the output electrodes is calculated as being low. A low impedance load is generally considered to be below 300 ohms. Both high and low impedance limits may be adjusted to match the instruments, processes, and/or procedures as necessary.

BRIEF DESCRIPTION OF THE DRAWINGS

10

Figure 1 presents an electrosurgical generator interfaced to a constant power control circuit having a current sampling circuit, linear conversion circuit and feedback correction circuit.

15

Figure 2 is the preferred embodiment of the linear conversion circuit shown in Figure 1.

Figure 3 is the preferred embodiment of the feedback correction circuit shown in Figure 1.

DETAILED DESCRIPTION OF THE INVENTION

20

For an electrosurgical generator 101 having a high voltage direct current (DC) supply 103 which is electrically connected to control an output switching radio frequency (RF) stage 105, a unique linear relationship exists between the control voltage supplied to the high voltage DC supply 103 and the root-mean-square (RMS) current generated by the electrosurgical generator 101. This unique linear relationship can be used to design a constant power control circuit 107 that functions as a feedback control loop to control the electrosurgical generator 101. The following mathematical derivations define this unique linear relationship.

30

It can be shown that:

$$V_{\text{control}} = V_{\text{dc}} / K_{\text{ps}};$$

where,

V_{control} = a control voltage supplied to the high voltage DC supply,

V_{dc} = the output voltage signal of the high voltage DC supply, and

K_{ps} = a feedback ratio of the high voltage DC supply.

It can further be shown that:

5 $V_{dc}^2 \times K_s = P_{out};$

where,

P_{out} = the output power of the electrosurgical generator 101, and

K_s = a linear constant (which can be empirically derived).

- 10 Therefore, the output power of the electrosurgical generator 101 is directly proportional to the square of the output voltage signal of the high voltage DC supply.

Thus, by substitution:

$$(V_{control} \times K_{ps})^2 \times K_s = P_{out}, \text{ or}$$

$$V_{control}^2 \times K_g = P_{out};$$

- 15 where,

$$K_g = K_{ps}^2 \times K_s.$$

Therefore, the output power of the electrosurgical generator 101 is proportional to the square of the control voltage supplied to the high voltage DC supply.

- 20 Examining the output of the generator we have:

$$P_{out} = V_{rms} \times I_{rms};$$

where,

V_{rms} = output RMS voltage of the electrosurgical generator 101, and

I_{rms} = output RMS current of the electrosurgical generator 101.

- 25

Accordingly, at a given load impedance = R:

$$V_{rms} = I_{rms} \times R,$$

and by substitution

$$P_{out} = I_{rms}^2 \times R.$$

- 30 By allowing R to equal a 'matched' load impedance we have

$$V_{control}^2 \times K_g = I_{rms}^2 \times R,$$

and therefore

$$V_{control}^2 = I_{rms}^2 \times R / K_g.$$

Consequently, for a given impedance $K_r = R / K_g$ the equation can be simplified to:

$$V_{\text{control}}^2 = I_{\text{rms}}^2 \times K_r.$$

Therefore, the square of the control voltage supplied to the high voltage DC supply
 5 103 is directly proportional to the square of the output RMS current of the
 electrosurgical generator 101. It can also be shown by similar derivation that the
 square of the control voltage supplied to the high voltage DC supply 103 is directly
 proportional to the square of the output RMS voltage of the electrosurgical generator
 101.

10 Thus, the above derivation implies that if either the output RMS current or
 voltage is sampled properly (I_{sample} & V_{sample} respectively) the control voltage supplied
 to the high voltage DC supply 103 may be used as a reference value in a feedback
 control loop to keep either the output RMS current or output RMS voltage constant.
 When the linear relationship of I_{rms} to I_{sample} is 'mapped' into the linear relationship of
 15 V_{control} to I_{rms} then a linear relationship can be derived between V_{control} and I_{sample} .
 When the scaling is done properly for a given power setting, V_{control} will equal I_{sample}
 at the 'matched' load impedance. Therefore, in a feedback circuit designed with the
 above mapping a feedback loop which keeps I_{sample} equal to V_{control} will by definition
 keep I_{rms} constant.

20 In accord with the above presented mathematical derivation, we have
 designed a constant power control circuit 107 for the electrosurgical generator 101,
 shown in Figure 1, having a power selection system 109 that produces a control
 voltage signal to control a high voltage direct current supply 103 which supplies a
 high voltage signal to an output switching radio frequency stage 105 thereby
 25 creating an electrosurgical energy between two output electrodes 111. The
 preferred electrosurgical generator 101 has a plurality of operational modes
 selectable within the power selection system 109, and a primary transformer winding
 113 within the output switching radio frequency stage 105, as shown in Figure 1.

The constant power control circuit 107, shown in Figure 1, includes a current
 30 sampling circuit 115, a linear conversion circuit 117 and a feedback correction circuit
 119.

In the preferred embodiment, the current sampling circuit 115 is inductively
 coupled to one of the output electrodes 111, as shown in Figure 1. Alternatively the

current sampling circuit 115 could be actively coupled, in circuit, with the output electrode.

The current sampling circuit 115 produces a sampled current signal that is proportional in amplitude to the average current flowing from the electrosurgical generator 101 through the one output electrode, an impedance load 121, and returning to the electrosurgical generator 101 through another output electrode.

The preferred embodiment of the current sampling circuit 115 includes an inductive coil element, similar in design and function to that of a secondary winding of a current transformer. Additional circuit elements function to transform the induced current into a proportional voltage signal and include a voltage drop resistor, a calibrating variable resistor, and elements that rectify and average the sampled current signal.

The current sampling circuit 115 supplies the sampled current signal to the linear conversion circuit 117. However, before the sampled current signal can be used as a feedback term, the mode crest factor for the selected electrosurgical generator 101 operational mode, needs to be compensated for. The linear conversion circuit 117, in Figures 1 and 2, compensates for the linear relationship between the sampled current signal and a 'true' sampled RMS value, which is of the form $I_{rms} = m \times I_{sample} + b$, where I_{rms} is a signal which is directly proportional to the RMS current, and m and b are given constants derived for a given crest factor. While electrosurgical generators 101 have a wide variety of different output wave shapes with varying crest factors, it is preferred that the crest factor for a given mode be significantly constant over a finite patient tissue impedance range, such as between 300 and 2500 ohms.

Accordingly, the linear conversion circuit 117 first multiplies the sampled current signal by the gain, m , and then adds the offset to it, b . When the values of m and b are chosen properly the resulting linear converted signal is directly proportional to the output RMS current of the electrosurgical generator 101. The preferred method for determining the proper values of m and b for a given operational mode and electrosurgical generator 101 includes collecting empirical data on the control voltage supplied to the high voltage DC supply 103 and the resulting output RMS current of the electrosurgical generator 101 and solving the linear equation, for m and b , by substitution.

The linear conversion circuit 117, shown in Figures 1 and 2, is electrically connected to the current sampling circuit 115. In the preferred embodiment, the linear conversion circuit 117 is also electrically connected to the power selection system 109 such that the operational mode of the electrosurgical generator 101 can be determined based on this connection. The linear conversion circuit 117 generates a linear converted signal and supplies this signal to the feedback correction circuit 119.

The preferred embodiment includes a linear multiplier generating means 201 within the linear conversion circuit 117, see Figure 2. The linear multiplier generating means 201 generates a plurality of unique multiplier reference signals (i.e., a factor 'm'). There is preferably one, unique, multiplier reference signal for each operational mode. The preferred embodiment, of the linear multiplier generating means 201 includes several resistive components connected to voltage sources, across which a predetermined voltage is maintained.

The preferred embodiment includes a linear offset generating means 203 within the linear conversion circuit 117, see Figure 2. The linear offset generating means 203 generates a plurality of unique offset reference signals (i.e., a factor 'b'). There is preferably one, unique, offset reference signal for each operational mode. The preferred embodiment of the linear offset generating means 203 includes several resistive components connected to voltage sources, across which a predetermined voltage is maintained.

The preferred embodiment also includes a plurality of multipliers 205, within the linear conversion circuit 117, see Figure 2. There is preferably one, corresponding, multiplier 205 for each operational mode. Each multiplier 205 is electrically connected to receive the sampled current signal and one unique multiplier reference signal from the linear multiplier generating means 201. Each multiplier 205 multiplies the sampled current signal and the unique multiplier reference signal associated with one operational mode to produce a unique multiplied signal for that operational mode. The preferred embodiment of the multiplier 205 includes a plurality of operational amplifiers.

The preferred embodiment includes a plurality of summers 207, within the linear conversion circuit 117, see Figure 2. There is preferably one, corresponding, summer 207 for each operational mode. Each summer 207 is electrically connected

to receive a unique multiplied signal and one unique offset reference signal from the linear offset generating means 203. Each summer 207 sums the offset reference signal associated with one operational mode and the unique multiplied signal associated with that operational mode to produce a unique linear converted signal
 5 for that operational mode. The preferred embodiment of the summer 207 includes configuring the plurality of operational amplifiers used as multipliers 205 to also function as summers 207.

The preferred embodiment includes a mode monitor 209, within the linear conversion circuit 117, see Figure 2. The mode monitor 209 is electrically
 10 connected to the power selection system 109, for identifying the operational mode of the electrosurgical generator 101 and producing an identified operational mode signal therefrom.

Closely associated with the mode monitor 209, is a signal selector 211 that is also within the linear conversion circuit 117, see Figure 2. The signal selector 211
 15 is electrically connected to receive the identified operational mode signal and the unique linear converted signal from each of the summers 207. The signal selector 211 selects the unique linear converted signal associated with the identified operational mode, and causes that linear converted signal to be supplied to the feedback correction circuit 119. In the preferred embodiment the mode monitor 209
 20 and signal selector 211 are embodied within a circuit including a digital processing component that activates and/or deactivates a plurality of electronic switching elements.

The feedback correction circuit 119, shown in Figures 1 and 3, is electrically connected to receive the linear converted signal from the linear conversion circuit
 25 117, the control voltage signal from the power selection system 109, and the voltage signal across the primary transformer winding 113. The feedback correction circuit 119 produces a feedback control signal and supplies the feedback control signal to the power selection system 109 so as to control the amount of electrosurgical energy created by the electrosurgical generator 101.

30 The feedback correction circuit 119 includes a subtractor 301, see Figure 3. The subtractor 301 is electrically connected to receive the linear converted signal from the linear conversion circuit 117 and the control voltage signal which is generated by the power selection system 109 and supplied to the high voltage DC

supply, see Figures 1 and 3. The subtractor 301 determines the difference in amplitude between the control voltage signal and the linear converted signal, and produces a delta signal proportional to the difference. The preferred embodiment of the subtractor 301 includes an operational amplifier component.

5 Also included in the feedback correction circuit 119 is an adder 303, see Figure 3. The adder 303 is electrically connected to receive the delta signal and the control voltage signal. The adder 303 adds the delta signal to the control voltage signal to produce the feedback control signal. The preferred embodiment includes an operational amplifier component.

10 Since holding the output RMS current constant for all impedances would be a physical impossibility based on the design limitations of the high voltage DC supply 103 and the output switching RF stage 105, it is preferred that the feedback control signal to the high voltage DC supply 103 be limited as a function of the impedance load 121 between the output electrodes 111.

15 In the preferred embodiment, the feedback correction circuit 119 includes a maximum control voltage reference generator 305 for generating a maximum control voltage reference signal, see Figure 3. The preferred embodiment uses an operational amplifier component connected to the control voltage signal to establish a maximum control voltage reference signal based thereon.

20 The maximum control voltage reference signal is supplied to a switcher 307 within the preferred feedback correction circuit 119, see Figure 3. The switcher 307 is also electrically connected to receive the feedback control signal from the adder 303. The switcher 307 substitutes the maximum control voltage reference signal for the feedback control signal when the feedback control signal is greater in
25 amplitude than the maximum control voltage reference signal, thereby limiting the electrosurgical generator's 101 output current through the output electrodes 111 when the impedance load 121 is at a low impedance level. The preferred embodiment of the switcher 307 includes an AND circuit created with diodes that passes the lower of the two signals as the feedback control signal.

30 When the impedance load 121 between the output electrodes 111 is high, the preferred constant power control circuit 107 should limit the output voltage of the electrosurgical generator 101 so as protect the electrosurgical generator 101, and reduce leakage currents and exit sparking.

In the preferred embodiment, the feedback correction circuit 119 shown in Figure 3, includes a high impedance reference generator 309 for generating a high impedance reference signal. The high impedance reference generator 309 is electrically connected to receive the control voltage signal. The preferred high impedance reference generator 309 establishes the high impedance reference signal by linearly converting the control voltage signal with an operational amplifier.

In the preferred embodiment a connector 311 is used for electrically connecting a comparator 313, within the feedback correction circuit 119, to the primary transformer winding 113, see Figures 1 and 3. The connector 311 provides the comparator 313 with the voltage across the primary transformer winding 113. The comparator 313 is also electrically connected to receive the high impedance reference signal. The comparator 313 compares the amplitude of the high impedance reference signal to the voltage across the primary transformer winding 113 and produces a high impedance detection signal that indicates the results of this comparison. In the preferred embodiment the comparator 313 includes an operational amplifier component.

The high impedance detection signal is received by a reducer 315, shown in Figure 3 of the preferred embodiment, which is electrically connected to the comparator 313 and to the switcher 307. The reducer 315 reduces, to an internally generated preset reduced voltage level signal, the amplitude of the feedback control signal from the switcher 307 when the voltage across the primary transformer winding 113 is greater than the high impedance reference signal as indicated by the high impedance detection signal. In the preferred embodiment, the reducer 315 includes a logic driven switched circuit and an adjustable resistor providing a reduced voltage level signal. The reducer 315 supplies the resulting feedback control signal to the power selection system 109.

Associated with the constant power control circuit 107 is a method for maintaining a generally constant output power from an electrosurgical generator 101 having a power selection system 109 that produces a control voltage signal to control a high voltage direct current supply 103 which supplies a high voltage signal to an output switching radio frequency stage 105 thereby creating an electrosurgical energy between two output electrodes 111.

The method includes the steps of inductively coupling to one output electrode, sensing the current flowing through the output electrode 111 and producing a sampled current signal proportional to the average current flowing through the output electrode. The method then continues with the steps of
 5 generating a multiplier reference signal, generating an offset reference signal, multiplying the sampled current signal and the multiplier reference signal, and then summing the offset reference signal to the product to producing a linear converted signal.

The method continues with the steps of connecting to the control voltage
 10 signal from the power selection system 109, determining the difference in amplitude between the control voltage signal and the linear converted signal, adding the difference determined by the subtraction means to the control voltage signal to produce a feedback control signal, and then supplying the feedback control signal to the power selection system 109 to control the amount of electrosurgical energy
 15 created.

To protect the electrosurgical generator 101 and the patient when the impedance load 121 is high, the method can include the steps of generating a high impedance reference signal, connecting to the primary transformer winding 113, comparing the amplitude of the high impedance reference signal to the voltage
 20 across the primary transformer winding 113, and reducing the amplitude of the feedback control signal when the voltage across the primary transformer winding 113 is greater than the high impedance reference signal.

To protect the electrosurgical generator 101 and patient when the impedance load 121 is low, the method can include the steps of generating a maximum control
 25 voltage reference signal and substituting the maximum control voltage reference signal for the feedback control signal when the feedback control signal is greater in amplitude than the maximum control voltage reference signal.

For electrosurgical generators 101 having a plurality of operational modes, the method can be modified to include the steps of generating a plurality of unique linear
 30 multiplier reference signals, one for each operational mode, and generating a plurality of unique linear offset reference signals, one for each operational mode. The method would then include the steps of multiplying the sampled current signal, separately and concurrently, with each of the unique multiplier reference signals to produce a

plurality of unique multiplied signals, one for each operational mode, and then summing each of the unique multiplied signals with the offset reference signal associated with the same operational mode to produce a plurality of unique linear converted signals, one for each operational mode. The method would continue with

5 the steps of connecting to the power selection system 109 to identify the operational mode selected, selecting the unique linear converted signal that matches the identified operational mode, and then causing that linear converted signal to be supplied to the feedback correction circuit 119.

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What is claimed is:

1. A constant power control circuit for an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes, the constant power control circuit comprising:

a current sampling circuit inductively coupled to one output electrode,
the current sampling circuit producing a sampled current signal proportional to the
10 average current flowing through the output electrode;

a linear conversion circuit electrically connected to the current sampling circuit for generating a linear converted signal, wherein the linear conversion circuit includes:

a linear multiplier generating means for generating a multiplier
15 reference signal;

a linear offset generating means for generating an offset reference signal;

a multiplier, electrically connected to the linear multiplier generating means, for multiplying the sampled current signal and the multiplier reference signal to produce a multiplied signal; and

a summer, electrically connected to the linear offset generating means, for summing the offset reference signal and the multiplied signal to produce the linear converted signal; and

25 a feedback correction circuit electrically connected to receive the linear converted signal from the linear conversion circuit and the control voltage signal from the power selection system for producing a feedback control signal which is supplied to the power selection system to control the amount of electrosurgical energy created, wherein the feedback correction circuit includes:

a subtractor, electrically connected to receive the linear converted
30 signal and the control voltage signal, for determining a difference in amplitude
between the control voltage signal and the linear converted signal and producing a
delta signal proportional to the difference; and

an adder, electrically connected to receive the delta signal and the control voltage signal, for adding the delta signal to the control voltage signal and producing the feedback control signal.

2. The constant power control circuit for an electrosurgical generator
 5 having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 1, wherein the output switching radio frequency stage has a primary transformer winding, and wherein the feedback correction circuit
 10 further includes:

a high impedance reference generator, electrically connected to receive the control voltage signal, for generating a high impedance reference signal;

a connector for electrically connecting the feedback correction circuit to the primary transformer winding;

- 15 a comparator, electrically connected to the connector and to receive the high impedance signal, for comparing the amplitude of the high impedance reference signal to the voltage across the primary transformer winding and producing a high impedance detection signal; and

a reducer, electrically connected to receive the high impedance
 20 detection signal and the feedback control signal, for reducing the amplitude of the feedback control signal when the high impedance detection signal indicates that the voltage across the primary transformer winding is greater than the high impedance reference signal.

3. The constant power control circuit for an electrosurgical generator
 25 having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 1, wherein the feedback correction circuit further includes:

30 a maximum control voltage reference generator, electrically connected to receive the control voltage signal, for generating a maximum control voltage reference signal; and

a switcher, electrically connected to receive the maximum control voltage reference signal and the feedback control signal, for substituting the maximum control voltage reference signal for the feedback control signal when the feedback control signal is greater in amplitude than the maximum control voltage reference signal.

4. The constant power control circuit for an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 1, wherein the power selection system has a plurality of operational modes and wherein the linear multiplier generating means generates a unique multiplier reference signal for each operational mode, and the linear offset generating means generates a unique offset reference signal for each operational mode, and wherein the linear conversion circuit further includes:

a plurality of multipliers, wherein there is one multiplier for each operational mode and each multiplier multiplies the sampled current signal and the unique multiplier reference signal associated with one operational mode to produce a unique multiplied signal for that operational mode;

a plurality of summers, wherein there is one summer for each operational mode and each summer sums the offset reference signal associated with one operational mode and the unique multiplied signal associated with that operational mode to produce a unique linear converted signal for that operational mode;

a mode monitor, electrically connected to the power selection system, for identifying the operational mode and producing an identified operational mode signal;

a signal selector, electrically connected to receive the identified operational mode signal and the unique linear converted signal from each of the summers, for selecting the unique linear converted signal associated with the identified operational mode, and causing that linear converted signal to be supplied to the feedback correction circuit.

5. A constant power control circuit for an electrosurgical generator having a power selection system that produces a control voltage signal to control a high

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the current sampling circuit producing a sampled current signal proportional to the average current flowing through the output electrode;

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multiplier reference signals, wherein there is one multiplier reference signal for each operational mode;

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operational mode signal and the unique linear converted signal from each of the

summers, for selecting the unique linear converted signal associated with the identified operational mode, and causing that linear converted signal to be supplied to the feedback correction circuit; and

5 a feedback correction circuit electrically connected to receive the linear converted signal from the linear conversion circuit and the control voltage signal from the power selection system for producing a feedback control signal which is supplied to the power selection system to control the amount of electrosurgical energy created, wherein the feedback correction circuit includes:

10 a subtractor, electrically connected to receive the linear converted signal and the control voltage signal, for determining a difference in amplitude between the control voltage signal and the linear converted signal and producing a delta signal proportional to the difference;

15 an adder, electrically connected to receive the delta signal and the control voltage signal, for adding the delta signal to the control voltage signal and producing the feedback control signal;

a maximum control voltage reference generator, electrically connected to receive the control voltage signal, for generating a maximum control voltage reference signal;

20 a switcher, electrically connected to receive the maximum control voltage reference signal and the feedback control signal, for substituting the maximum control voltage reference signal for the feedback control signal when the feedback control signal is greater in amplitude than the maximum control voltage reference signal

25 a high impedance reference generator, electrically connected to receive the control voltage signal, for generating a high impedance reference signal;

a connector for electrically connecting the feedback correction circuit to the primary transformer winding;

30 a comparator, electrically connected to the connector and to receive the high impedance signal, for comparing the amplitude of the high impedance reference signal to the voltage across the primary transformer winding and producing a high impedance detection signal; and

a reducer, electrically connected to receive the high impedance detection signal and the feedback control signal, for reducing the amplitude of the

feedback control signal when the high impedance detection signal indicates that the voltage across the primary transformer winding is greater than the high impedance reference signal.

6. A method for maintaining a generally constant output power from an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes, the method including the steps of:
 - inductively coupling to one output electrode;
 - sensing the current flowing through the output electrode;
 - producing a sampled current signal proportional to the average current flowing through the output electrode;
 - generating a multiplier reference signal;
 - generating an offset reference signal;
 - multiplying the sampled current signal and the multiplier reference signal; then
 - summing the offset reference signal to the product to producing a linear converted signal;
 - connecting to the control voltage signal from the power selection system;
 - determining the difference in amplitude between the control voltage signal and the linear converted signal;
 - adding the difference determined by the subtraction means to the control voltage signal to produce a feedback control signal.
 - supplying the feedback control signal to the power selection system to control the amount of electrosurgical energy created.

7. The method for maintaining a generally constant output power from an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 6, wherein the output

switching radio frequency stage has a primary transformer winding, and wherein the method further includes the steps of:

- generating a high impedance reference signal;
- connecting to the primary transformer winding;
- 5 comparing the amplitude of the high impedance reference signal to the voltage across the primary transformer winding; and
- reducing the amplitude of the feedback control signal when the voltage across the primary transformer winding is greater than the high impedance reference signal.

- 10 8. The method for maintaining a generally constant output power from an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 6, the method further
- 15 including the steps of:

- generating a maximum control voltage reference signal; and
- substituting the maximum control voltage reference signal for the feedback control signal when the feedback control signal is greater in amplitude than the maximum control voltage reference signal.

- 20 9. The method for maintaining a generally constant output power from an electrosurgical generator having a power selection system that produces a control voltage signal to control a high voltage direct current supply which supplies a high voltage signal to an output switching radio frequency stage thereby creating an electrosurgical energy between two output electrodes of Claim 6, wherein the
- 25 power selection system has a plurality of operational modes and wherein the method further includes the steps of:

- generating a plurality of unique linear multiplier reference signals, one for each operational mode;

- generating a plurality of unique linear offset reference signals, one for
- 30 each operational mode;

- multiplying the sampled current signal, separately and concurrently, with each of the unique multiplier reference signals to produce a plurality of unique multiplied signals, one for each operational mode;

summing each of the unique multiplied signals with the offset reference signal associated with the same operational mode to produce a plurality of unique linear converted signals, one for each operational mode;

connecting to the power selection system to identify the operational
5 mode selected;

selecting the unique linear converted signal that matches the identified operational mode; and then

causing that linear converted signal to be supplied to the feedback correction circuit.

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ELECTROSURGICAL GENERATOR POWER CONTROL CIRCUIT AND METHOD

5 A constant power control circuit for an electrosurgical generator and a method for maintaining the electrical power output of an electrosurgical generator at a generally constant value throughout a given tissue impedance range are disclosed. The constant power control circuit and the method recognize and use the unique and simple linear characteristics associated with certain electrosurgical generator designs to monitor and control the electrical power output without having to calculate or monitor the actual output power. The constant power control circuit includes a
10 current sampling circuit, a linear conversion circuit, and a feedback correction circuit. The constant power control circuit may also include protection circuitry that prevents the electrosurgical generator from being over-driven during high and/or low impedance loading, and reduces the severity of exit sparking by providing a quick response to high impedance indications while nonetheless maintaining increased
15 power levels throughout a preset, nominal impedance range. The constant power control circuit and method may be included as an integral part of the overall electrosurgical generator's circuitry, or may be embodied as a separate unit that connects to, and controls, an electrosurgical generator. The constant power control circuit and method may be embodied through a variety of analog and/or digital circuit
20 components or arrangements, including software running on computational and memory circuitry.

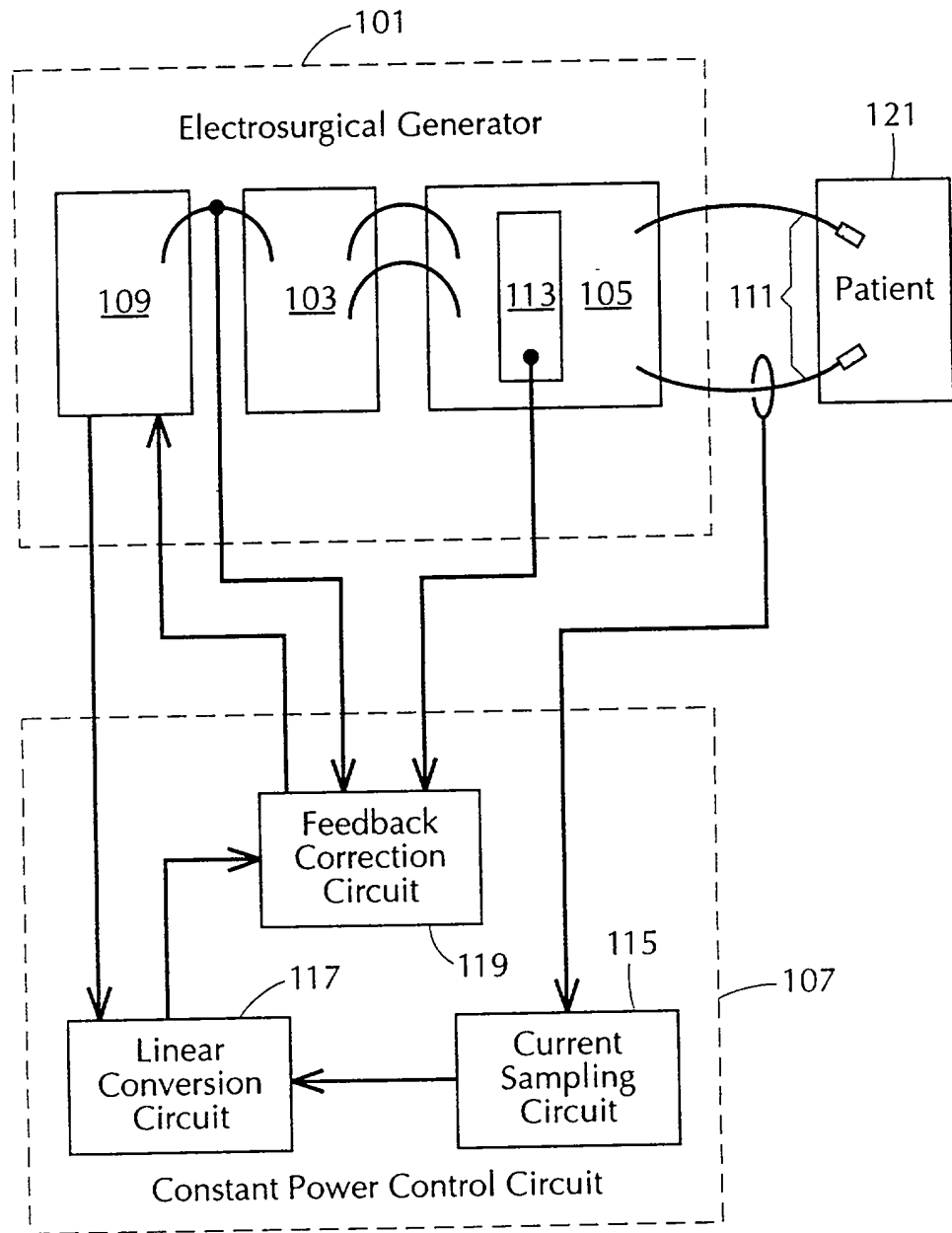
FIG. 1

FIG. 2

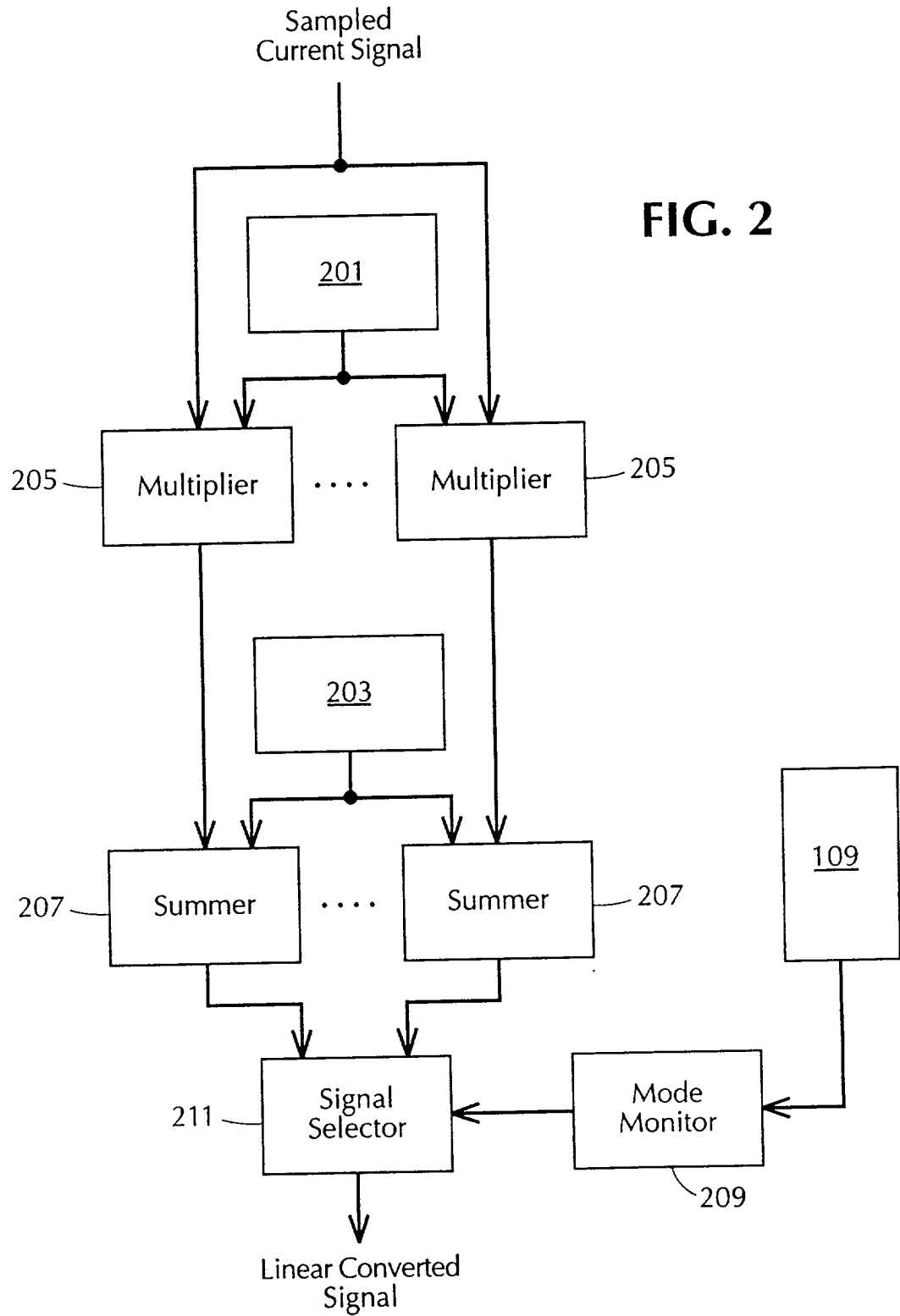
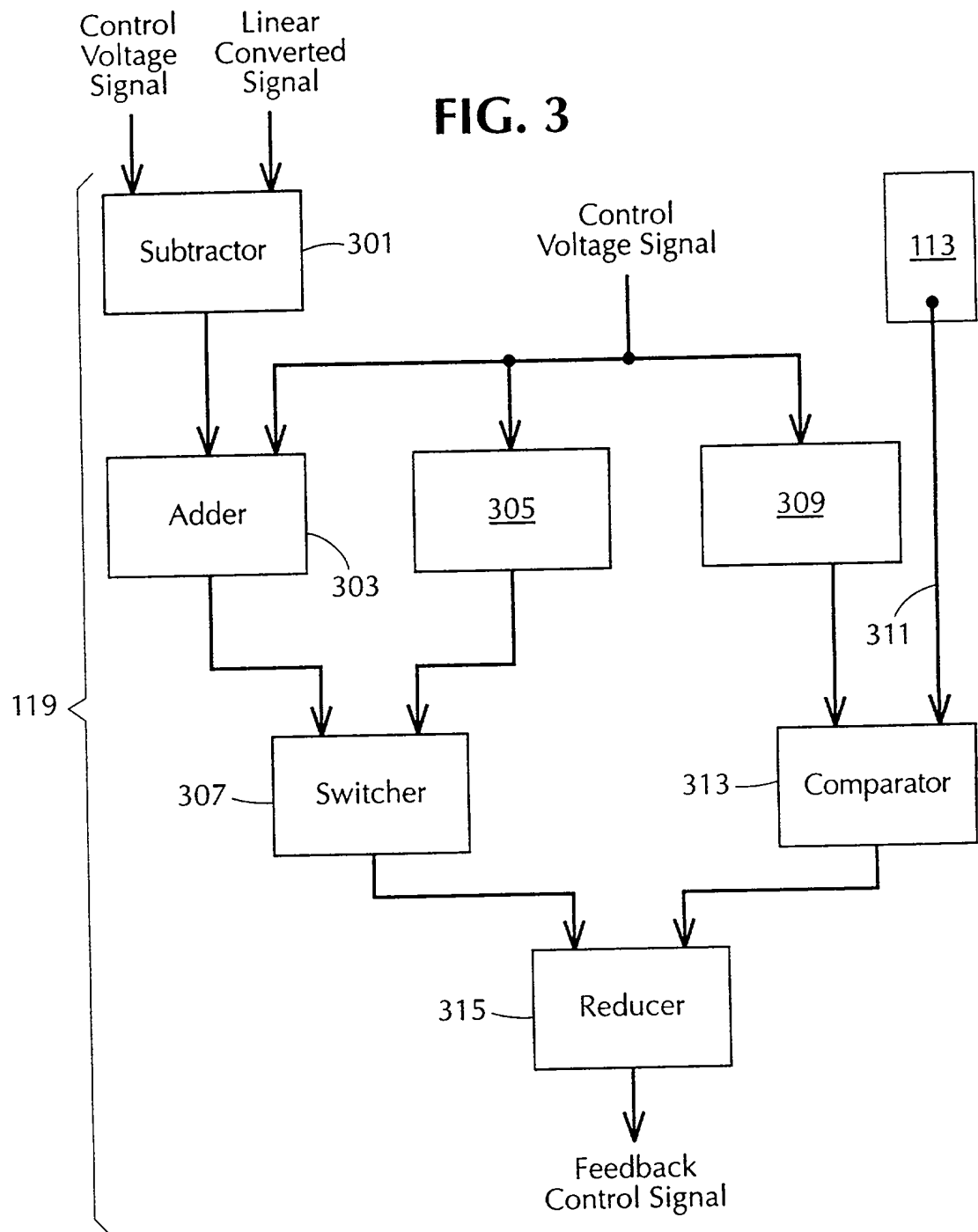


FIG. 3



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COMBINED DECLARATION FOR PATENT APPLICATION AND POWER OF ATTORNEY
(Includes Reference to PCT International Applications) Page 2 of 4

Docket Number
PC8656ALP

I hereby claim the benefit under Title 35, United States Code, §120 of any United States application(s) or PCT international application(s) designating the United States of America that is/are listed below and, insofar as the subject matter of each of the claims of this application is not disclosed in that/those prior application(s) in the manner provided by the first paragraph of Title 35, United States Code, §112, I acknowledge the duty to disclose material information as defined in Title 37, Code of Federal Regulations, §1.56(a) which occurred between the filing date of the prior application(s) and the national or PCT international filing date of this application.

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POWER OF ATTORNEY: As a named inventor, I hereby appoint the following attorney(s) and/or agent(s) to prosecute this application and transact all business in the Patent and Trademark Office connected therewith. (List name and registration number)

SEE ATTACHED LIST

Send Correspondence to: Dr. Peter C. Richardson,
PFIZER INC, 235 East 42nd Street,
New York, NY 10017-5755, U.S.A.

Direct Telephone Calls To:
Aaron Passman (303) 581-6614

201	FULL NAME OF INVENTOR	BECKER	DANIEL	J.
	RESIDENCE & CITIZENSHIP	BROOMFIELD	COLORADO	UNITED STATES OF AMERICA
	POST OFFICE ADDRESS	1225 W. 7TH AVENUE	BROOMFIELD	COLORADO/80020/U.S.A.
202	FULL NAME OF INVENTOR	KLICEK	MICHAEL	STEVE
	RESIDENCE & CITIZENSHIP	TROY	MICHIGAN	UNITED STATES OF AMERICA
	POST OFFICE ADDRESS	2964 ROUNDTREE DRIVE	TROY	MICHIGAN/48083/U.S.A.
203	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
	POST OFFICE ADDRESS	POST OFFICE ADDRESS	CITY	STATE/ZIP CODE/COUNTRY
204	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
	POST OFFICE ADDRESS	POST OFFICE ADDRESS	CITY	STATE/ZIP CODE/COUNTRY

205	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
	POST OFFICE ADDRESS	POST OFFICE ADDRESS	CITY	STATE/ZIP CODE/COUNTRY
206	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
	POST OFFICE ADDRESS	POST OFFICE ADDRESS	CITY	STATE/ZIP CODE/COUNTRY
207	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
	POST OFFICE ADDRESS	POST OFFICE ADDRESS	CITY	STATE/ZIP CODE/COUNTRY
208	FULL NAME OF INVENTOR	FAMILY NAME	FIRST GIVEN NAME	SECOND GIVEN NAME
	RESIDENCE & CITIZENSHIP	CITY	STATE OR FOREIGN COUNTRY	COUNTRY OF CITIZENSHIP
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I hereby declare that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under section 1001 of Title 18 of the United States Code, and that such willful false statements may jeopardize the validity of the application or any patent issuing thereon.

SIGNATURE OF INVENTOR (201) <i>David J. Buel</i>	SIGNATURE OF INVENTOR (202) <i>Michael S. Kline</i>	SIGNATURE OF INVENTOR (203)
DATE <i>9/26/95</i>	DATE <i>9/14/95</i>	DATE
SIGNATURE OF INVENTOR (204)	SIGNATURE OF INVENTOR (205)	SIGNATURE OF INVENTOR (206)
DATE	DATE	DATE
SIGNATURE OF INVENTOR (207)	SIGNATURE OF INVENTOR (208)	SIGNATURE OF INVENTOR (209)
DATE	DATE	DATE

POWER OF ATTORNEY (CONTINUED)

PETER C. RICHARDSON, Reg. No. 27,526; MICHAEL J. PANTULIANO, Reg. No. 18,971; ALLEN J. SPIEGEL, Reg. No. 25,749; AARON PASSMAN, Reg. No. 26,783; GEZINA HOLTRUST, Reg. No. 28,222; J. TREVOR LUMB, Reg. No. 28,567; LAWRENCE C. AKERS, Reg. No. 28,587; RAYMOND W. AUGUSTIN, Reg. No. 28,588; SEYMOUR G. BEKELNITZKY, Reg. No. 28,589; PAUL H. GINSBURG, Reg. No. 28,718; MARK DRYER, Reg. No. 28,775; ELIZABETH O. SLADE, Reg. No. 29,011; JOHN L. LAPIERRE, Reg. No. 29,185; ERIC M. LEE, Reg. No. 30,471; JAMES T. JONES, Reg. No. 30,561; GREGG C. BENSON, Reg. No. 30,997; A. DEAN OLSON, Reg. No. 31,185; ROBERT F. SHEYKA, Reg. No. 31,304; HOWARD R. JAEGER, Reg. No. 31,376; GROVER F. FULLER, JR., Reg. No. 31,760; MERVIN E. BROKKE, Reg. No. 32,723; KAREN DEBENEDICTIS, Reg. No. 32,977; VALERIE M. FEDOWICH, Reg. No. 33,688; PHILIP C. STRASSBURGER, Reg. No. 34,258; BRYAN C. ZIELINSKI, Reg. No. 34,462; ROBERT T. RONA, Reg. No. 36,257; and GARTH BUTTERFIELD, Reg. No. 36,997.